11

REPRESENTATION OF LINEAR DIGITAL NETWORKS



Figure Sll.1-1

We have drawn the flow-graph to graphically correspond closely to the block diagram in (a). There are, of course, many other ways of drawing the flow-graph, for example



Figure Sll.1-2

Solution 11.2

(a) The equations corresponding to this flowgraph are: $W_1(z) = X(z)$ $W_2(z) = W_1(z) + z^{-1} W_4(z)$ $W_{3}(z) = 2 W_{2}(z)$ $W_4(z) = 2 W_1(z) + 3 W_3(z)$ $Y(z) = W_3(z)$ or, in matrix form $\begin{vmatrix} w_1 \\ w_2 \\ w_3 \end{vmatrix} = \begin{vmatrix} 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & z^{-1} \\ 0 & 2 & 0 & 0 \end{vmatrix} \begin{vmatrix} w_1 \\ w_2 \\ w_3 \\ w_3 \end{vmatrix} + \begin{vmatrix} 0 \\ 0 \\ 0 \\ 0 \end{vmatrix} X(z)$ $Y(z) = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix}$ ^w1 ^W2 ^W3 $\mathbf{F}_{c}^{t} = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 2 & 0 & 0 \\ 2 & 0 & 3 & 0 \end{bmatrix}$ and $\mathbf{F}_{d}^{t} = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}$

Solution 11.3

Let us carry this out by obtaining the transfer function for each of the networks. For network 1: $Y(z) = 2r \cos \theta z^{-1} Y(z) - r^2 z^{-2} Y(z) + X(z)$ or

$$H_{1}(z) = 1/[1 - 2r \cos \theta z^{-1} + r^{2} z^{-2}]$$

For network 2:

Define $W_1(z)$ as shown in Figure Sll.3-1.



Figure Sll.3-1

then

$$\begin{split} & \mathbb{W}_{1}(z) = \mathbb{X}(z) - r \sin \theta \ z^{-1} \ \mathbb{Y}(z) + r \cos \theta \ z^{-1} \ \mathbb{W}_{1}(z) \\ & \mathbb{Y}(z) = r \sin \theta \ z^{-1} \ \mathbb{W}_{1}(z) + r \ \cos \theta \ z^{-1} \ \mathbb{Y}(z) \\ & \text{Solving for } \mathbb{Y}(z) \text{ in terms of } \mathbb{X}(z) \text{ we obtain} \\ & \mathbb{Y}(z) = \mathbb{X}(z) \ r(\sin \theta) \ z^{-1} / \lceil 1 - 2r \ \cos \theta \ z^{-1} + r^{2} \ z^{-2} \rceil \end{split}$$

or

$$H_{2}(z) = \frac{r(\sin \theta) z^{-1}}{[1 - 2r \cos \theta z^{-1} + r^{2} z^{-2}]}$$

Thus both networks have the same poles.

Solution 11.4

With the nodes ordered as shown in the figure, the matrix \underline{F}_{c}^{t} is

 $\underline{\mathbf{F}}_{\mathbf{C}}^{\mathbf{t}} = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 1 & 0 & 1 & 0 \end{bmatrix}$

Thus with the nodes arranged in the order 1-2-3-4 they can be computed in sequence since the matrix \underline{F}_{c}^{t} is zero on and above the main diagonal. There is no other ordering possible.

Solution 11.5

(a) A flow-graph in terms of H_1 , H_2 , H_3 and H_4 can be drawn as



Figure Sll.5-1

However we want to draw the flow-graph using branch transmittances which are constant or a constant times z^{-1} . Thus we replace H_1 , H_2 , H_3 and H_4 by their flow-graph implementations to obtain



Figure Sll.5-2

(b) We now want to connect a network $H_B(z)$ to the right-hand side of the above network. Observe that there is a delay-free path from X_2 to Y_2 . Consequently, if the system $H_B(z)$ has a delay-free path from its input to its output, the total system will have a delay-free loop and thus will be noncomputable. By contrast, if $H_B(z)$ does not have a delay-free path from its input to its output, the overall system will be computable. A necessary and sufficient condition such that $H_B(z)$ is not delay-free from input to output is that $h_B(n)$, its unit sample

response be zero at n = 0 (we are, of course, assuming that the system is causal.) This is guaranteed if $H_B(z)$ can be written in the form

 $H_B(z) = z^{-1} \hat{H}_B(z)$ where $\hat{H}_B(z)$ is also causal, or equivalently that $\lim_{z\to\infty} H_B(z) = 0.$ Resource: Digital Signal Processing Prof. Alan V. Oppenheim

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